

Maxim > Design Support > Technical Documents > Application Notes > Audio Circuits > APP 4504 Maxim > Design Support > Technical Documents > Application Notes > Signal Generation Circuits > APP 4504

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## **APPLICATION NOTE 4504**

## Locked-Sync Sine Generator Covers Three Decades with Low Distortion

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Abstract: This circuit synchronizes a sinewave output through three decades of frequency (20Hz to 20kHz) while maintaining low THD and constant amplitude.

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Analog applications such as testing, calibration, and general system operation often require a sine waveform of accurate amplitude and frequency, with low total harmonic distortion (THD). Some applications demand that the generator of such waveforms have the ability to accurately synchronize the output with an external timing signal. Simple sinewave generators can offer various degrees of this performance, but maintaining low THD with constant amplitude is a problem, particularly if



the output and the synchronization signal must remain locked through an extended range of frequencies.

The circuit of **Figure 1** can synchronize a sinewave output through three decades of frequency (20Hz to 20kHz) while maintaining low THD and constant amplitude (**Table 1**). The synchronizer IC (74HC4046) is a phase-locked loop (PLL) complete with VCO and phase/frequency detector. It has three internal phase detectors, but the one used here (number 2) has a frequency-capture range equal to that of the VCO frequency range ( $f_{MAX} - f_{MIN}$ ).



Figure 1. This 3-IC sinewave generator covers three frequency decades, provides low distortion, and can be synched to an external signal.

The circuit's general-purpose binary frequency divider (the 74HC4060) is connected between the VCO output and the 74HC4046 feedback input (the phase/frequency comparator input), and set for a division ratio of 64. When the PLL is locked, therefore, the Q6 output of the 74HC4060 generates a square wave equal to  $1/64^{th}$  of the VCO output frequency. The components that determine the 74HC4046 center frequency (C1 and R1) are dimensioned so the VCO frequency can range from (20 × 64) to (20,000 × 64), from the minimum to the maximum level of the VCO's input-voltage range.

A switched-capacitor lowpass filter (MAX297), whose cut-off frequency (by design) equals  $1/50^{\text{th}}$  of the clock frequency applied to it, has for signal input the same square wave used for the PLL feedback, and its clock input is attached to the VCO output. Because the clock and signal inputs always have a frequency ratio of 64, the input signal always falls within the filter bandpass. No input harmonics fall within this bandpass, because the ratio of F<sub>CLOCK</sub> to frequency is < 50 for all of them. (For the lowest-possible (second) harmonic, the ratio is 32.)

The fact that the filter's input signal is a square wave (with 50% duty cycle) helps this application, because a square wave contains only odd harmonics of the fundamental, and the lowest-frequency harmonic is the third, which is well within the filter's deep-attenuation range. Output amplitude and distortion over the frequency range are shown in Table 1.

You can frequency-modulate the synchronization signal, but that entails a compromise between the synchronization tracking speed (or maximum modulation frequency and depth) and the frequency locking range, which is set by the PLL's lowpass filter components (R2, R3, and C2). Modulation speed is limited for the values shown, because those values are optimized for an extended-frequency locking range. More information, including a full data sheet for the MAX297, are available for download.

Table 1. THD an	d Amplitu	ide vs. Frequency,	for	Figure	1	Circuit
Frequency (Hz)	THD (%)	Amplitude (V <sub>RMS</sub> )				

riequency (nz)	THD (%)	
20	2.775	1.470
50	2.650	1.472

100	2.525	1.472
200	2.250	1.473
500	1.002	1.473
1000	0.186	1.473
2000	0.260	1.472
5000	0.330	1.473
10000	0.405	1.473
20000	0.022	1.472

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